

Systems of Linear Eq and Gaussian Elim (1A) (1B)

Vector - ordered list of numbers $\vec{x} = \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix}$

Matrix - rectangular array of numbers $\begin{bmatrix} A_{11} & \dots & A_{1n} \\ \vdots & & \vdots \\ A_{m1} & \dots & A_{mn} \end{bmatrix}$

Gaussian Elim - algorithm that reduces lin eq (can obtain unique sol, infinite sol, no sol)

2A) Vector Addition: Comm: $\vec{x} + \vec{y} = \vec{y} + \vec{x}$ add $\vec{x} + \vec{0} = \vec{x}$
 Assoc: $(\vec{x} + \vec{y}) + \vec{z} = \vec{x} + (\vec{y} + \vec{z})$ add inv $\vec{x} + (-\vec{x}) = \vec{0}$

For water pump: rows - represent proportion of water a reservoir will draw from other reservoirs
 columns - how water will be distributed between other reservoirs the next day

Matx vector form: $A\vec{x} = \vec{b}$

3) Linear dependence - redundancy - one vector can be written as linear comb of another
 Def: vectors $\{\vec{v}_1, \vec{v}_2, \dots, \vec{v}_n\}$ are linearly dependent if $\alpha_1, \dots, \alpha_n$ such that $\alpha_1 \vec{v}_1 + \alpha_2 \vec{v}_2 + \dots + \alpha_n \vec{v}_n = \vec{0}$ and $\alpha_n \neq 0$

Linear independence - not linearly dependent $\alpha_n = 0$

Span - span of vectors $\{\vec{v}_1, \dots, \vec{v}_n\}$ is set of all their comb of $\{\vec{v}_1, \dots, \vec{v}_n\}$

2B) Matx Mult: $\begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix} \begin{bmatrix} b_{11} & b_{12} \\ b_{21} & b_{22} \end{bmatrix} = \begin{bmatrix} a_{11}b_{11} + a_{12}b_{21} & a_{11}b_{12} + a_{12}b_{22} \\ a_{21}b_{11} + a_{22}b_{21} & a_{21}b_{12} + a_{22}b_{22} \end{bmatrix}$

Not commutative $AB \neq BA$ Is associative $(AB)C = A(BC)$

6) Equivalent for $n \times n$ A

- a) A is invertible
 - b) \Leftrightarrow eq $A\vec{x} = \vec{b}$ has unique sol for \vec{b}
 - c) \Leftrightarrow A has linearly independent cols (full rank)
 - d) \Leftrightarrow A has trivial null space ($\text{null}(A) = \{\vec{0}\}$)
 - e) $\Leftrightarrow \det(A) \neq 0$
- Probable
 a) A is invertible $\Rightarrow A\vec{x} = \vec{b}$ unique sol
 b) A is invertible \Rightarrow linearly independent
 c) A is invertible \Rightarrow trivial null space

7) Vector space (V) set of vectors and two operators satisfy

- Vector addition
- Associative $\vec{u} + (\vec{v} + \vec{w}) = (\vec{u} + \vec{v}) + \vec{w}$
 - Comm $\vec{u} + \vec{v} = \vec{v} + \vec{u}$
 - additive $\vec{v} + \vec{0} = \vec{v}$ and $\vec{v} + (-\vec{v}) = \vec{0}$
 - Closure sum $\vec{u} + \vec{v}$ must be in V
- Scalar mult
- Associative $\alpha(\beta\vec{v}) = (\alpha\beta)\vec{v}$
 - Multi Identity $1 \cdot \vec{v} = \vec{v}$
 - Distributive $\alpha(\vec{u} + \vec{v}) = \alpha\vec{u} + \alpha\vec{v} = (\alpha + \beta)\vec{v} = \alpha\vec{v} + \beta\vec{v}$
 - Closure any vector \vec{v} scalar α $\alpha\vec{v}$ must be in V

Basis - of vector space V (min set of vects to represent all vectors)

1) $\vec{v}_1, \vec{v}_2, \dots, \vec{v}_n$ linearly independent

2) scalars such that $\vec{v} = \alpha_1 \vec{v}_1 + \alpha_2 \vec{v}_2 \dots$

Dimension of a vector space is number of basis vectors

⑧ **Subspace** - (W, F) is a subspace if W is a subset of V and (W, F) is a vector space

Nullspace (A) := solutions to $A\vec{x} = \vec{0}$

Column space of (A) = span of the columns of A

rank - max num of linearly independent columns

Proofs

① Write what you know and what you want to prove

② - Try small simple example to see if you find patterns

③ manipulate both sides and simplify Justify each step

④ 1) Constructive proofs (generate example/solution)

2) proofs by contradiction

Knows: $A\vec{v} = \lambda\vec{v}$ Subset $A \subseteq B$

Ex) Does inverse exist? many column's are linearly independent, invertible, exact same # rows as columns

Linear independence: $\mathbb{R}^4/\mathbb{K} = (\mathbb{R}_2 + \mathbb{R}_3)\mathbb{K}$

Basis: The basis for null (A) is $\left\{ \begin{bmatrix} 1 \\ -1 \\ 0.5 \end{bmatrix} \right\}$

non trivial nullspace \Rightarrow linearly dependent \Rightarrow no unique sol.

Thm: If $A\vec{x} = \vec{b}$ has two or more sol, then col of A are linearly dependent

let \vec{x}_1, \vec{x}_2 be two sol

$$A\vec{x}_1 = \vec{b} \quad A\vec{x}_2 = \vec{b}$$

$$A(\vec{x}_1 - \vec{x}_2) = \vec{0}$$

$$\vec{y} = \vec{x}_1 - \vec{x}_2 \neq \vec{0}$$

$$A \begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_n \end{bmatrix} = \vec{0}$$

$$y_1 \vec{a}_1 + y_2 \vec{a}_2 + \dots + y_n \vec{a}_n = \vec{0}$$

WLOG let $y_1 \neq 0$

$$\vec{a}_1 = -\frac{y_2}{y_1} \vec{a}_2 + \dots - \frac{y_n}{y_1} \vec{a}_n$$

show $c_1 \vec{a}_1 + c_2 \vec{a}_2 + \dots + c_n \vec{a}_n = \vec{0}$

Given $H\vec{\psi}_1 = e\vec{T}_1 \quad H\vec{\psi}_2 = e\vec{T}_2$

$$\vec{\psi} = \alpha_1 \psi_1 + \alpha_2 \psi_2$$

$$H\vec{\psi} = H(\alpha_1 \psi_1 + \alpha_2 \psi_2) = H\alpha_1 \psi_1 + H\alpha_2 \psi_2 = \alpha_1 e\vec{T}_1 + \alpha_2 e\vec{T}_2 = e(\alpha_1 \vec{T}_1 + \alpha_2 \vec{T}_2) = e\vec{T}$$

Circuits: Resistors

- Kirchhoff Current Law (KCL) - All currents entering a node equal the sum of all currents exiting that node
- Kirchhoff Voltage Law (KVL) - Voltage around all loops sum to 0

Ohm's Law

$$V = IR \quad R = \frac{V}{I} \quad I = \frac{V}{R}$$

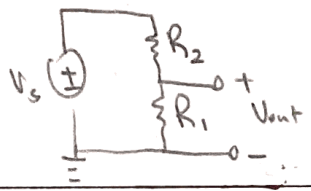
Resistance

$$R = \rho \frac{L}{A}$$

ρ ← resistivity
 L ← length
 A ← area of cross section

Voltage Divider

$$V_{out} = V_s \left(\frac{R_1}{R_1 + R_2} \right)$$



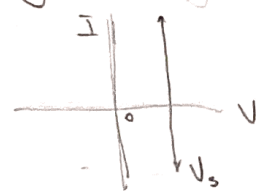
Current Divider

$$I_n = I_s \frac{R_{rest}}{R_n + R_{rest}}$$

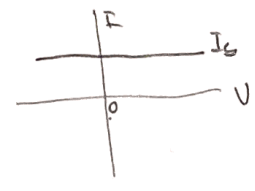


I/V

Voltage Source



Current Source



Passive Sign Convention:



Open circuit means $i = 0$

- Can combine 3+ in series but not 3+ in parallel

Power / Energy

$$P = \frac{dE}{dt} = V \cdot I = \frac{V^2}{R}$$

$$E = P \cdot \Delta t = \int_0^t P(t) dt$$

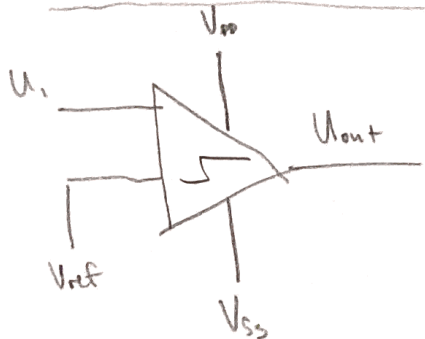
My Power dissipates Energy

Energy [J] or [W·s]

$$P = I^2 R$$

Power [W]

Op-Amp (Comparator)



if $u_i > V_{ref} \Rightarrow u_{out} = V_{DD}$

if $u_i < V_{ref} \Rightarrow u_{out} = V_{SS}$

$$V_{out} = A(V^+ - V^-)$$

$i = 0$

Superposition

- Set other independent sources to null
- Find each $V_{out, k}$
- Sum all $V_{out, sources}$ to find V_{out}

Capacitance

$$Q_c = C \cdot V_c$$

↑ Charge [C] ↑ Capacitance [F]

$$C = \frac{A}{d} \epsilon$$

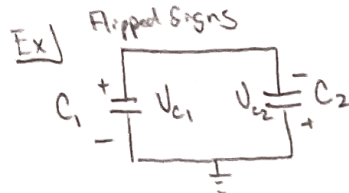
↑ permittivity

$$E = \frac{1}{2} C V_c^2$$

Charging: $I(t) = C \frac{dV}{dt}$

Discharging: $-I(t) = C \frac{dV}{dt}$

$$V_c(t) = \frac{I_c}{C} \cdot T + V_0$$



$$Q_{total} = Q_{c1} + Q_{c2}$$

$$Q = C_1 V_{c1} + (-C_2 V_{c2})$$

Conservation of Charge

$Q[t] = Q[t+1]$ - Charge in system remains same after time step

Thevenin

$$V_{Th} = V_{oc}$$

$$R_{Th} = R_{NOR}$$

$$V_{Th} = I_N \cdot R_{Th}$$

Thevenin is open circuit voltage between out ports

Way to solve:

- find V at A

- find V at B w/o taking into account A

- subtract

- Null Independent Source

- Calc R_{Th}

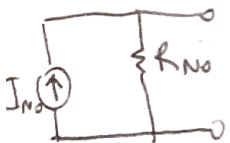
Norton

$$I_{No} = I_{meas}$$

Current flowing through output ports when ports are shorted

Way to solve:

- connect A & B by wire and ignore in between



Node Voltage Analysis (NVA)

- 1) Label Voltages at all Nodes
- 2) Label Element Voltages and Currents & Passive sign conv.
- 3) Write KCL Equations of Element Currents
- 4) Expressions for Element Currents using KVL
- 5) Substitute Expressions for Element Currents into KCL
- 6) Solve

Tips

- Redraw circuits if necessary
- Pay attention to units
- See what variable/terms answer is supposed to be in

Cross-Correlation

Def: measure of the similarity between two signals

$$\text{Corr}_{\vec{x}}[\vec{y}][\vec{K}] = \sum_{i=-\infty}^{\infty} x[i] y[i-k]$$

Because of noise we want $A\vec{x} = \vec{b}$ but to minimize noise we use $\vec{e} = \vec{b} - A\vec{x}$ minimize $\|\vec{e}\|^2$ find smallest error

Least Squares

Def: Solves approx. systems in the presence of noise

$$\vec{\hat{x}} = (A^T A)^{-1} A^T \vec{b} \quad \text{where } A \text{ is independent}$$

Note: Use for tall and thin / overdetermined matrices

Orthogonal Matching Pursuit (OMP)

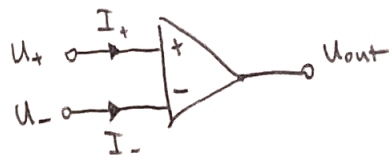
Def: unmixes signals from multiple beacons, uncover messages

Known: matrix M , vector \vec{y} , sparsity level k or threshold for norm

- ① Find inner product for each vector, \vec{m}_i ; with \vec{e} (first iter: $\vec{e} = \vec{y}$)
- ② Add vector that had max inner prod to matrix A
- ③ Use proj / least squares to compute $\vec{x} = (A^T A)^{-1} A^T \vec{y}$
- ④ Update Error vector $\vec{e} = \vec{y} - A\vec{x}$
- ↻ Repeat until reached sparsity level

- Linearize distance formula eqs

Op Amp



Golden Rules:

Ideal op Amp

- ① $I_+ = I_- = 0$
- ② $A \rightarrow \infty$

Ideal & Negative Feedback

- ① $U_+ = U_-$

Unity Gain Buffer

- ① $V_{in} = V_{out}$

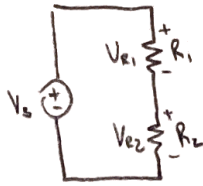
loading - small resistance causes it to draw a lot of current causing large voltage drop, make it harder bc it will behave differently depending on what it's connected to

- Buffers allow us to split into blocks to analyze separately

Design Problem

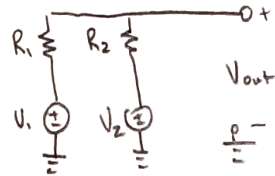
- ① Restate important goals for design
- ② Describe strategy using block diagram
- ③ Implement
- ④ Check implementation with Step 1

Voltage Divider



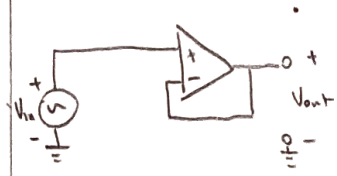
$$V_{R_2} = V_s \left(\frac{R_2}{R_1 + R_2} \right)$$

Voltage Summer



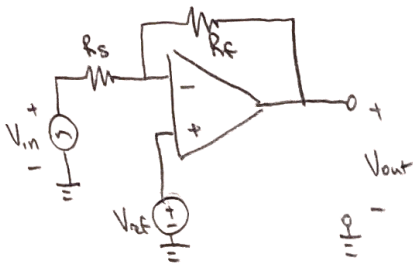
$$V_{out} = V_1 \left(\frac{R_2}{R_1 + R_2} \right) + V_2 \left(\frac{R_1}{R_1 + R_2} \right)$$

Unity Gain Buffer



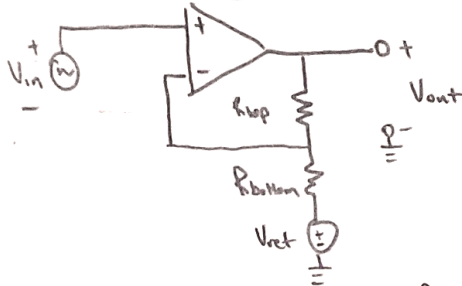
$$\frac{V_{out}}{V_{in}} = 1$$

Inverting Amplifier



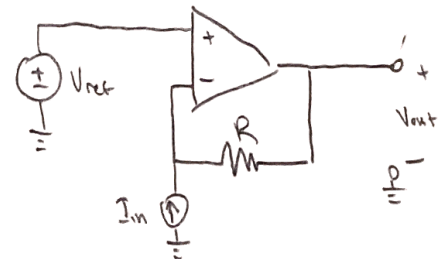
$$V_{out} = V_{in} \left(-\frac{R_f}{R_s} \right) + V_{ref} \left(\frac{R_f}{R_s} + 1 \right)$$

Non-inverting Amplifier



$$V_{out} = V_{in} \left(1 + \frac{R_{top}}{R_{bottom}} \right) - V_{ref} \left(\frac{R_{top}}{R_{bottom}} \right)$$

Transresistance Amplifier



$$V_{out} = i_{in} (-R) + V_{ref}$$

Orthonormal Matrix

Columns \vec{a}_i are

- ① Orthogonal $\langle a_i, a_j \rangle = 0$
- ② Normalized $\|a_i\| = 1$

Eigenvalues

$$A\vec{x} = \lambda\vec{x}$$

$$\det(A - \lambda I) = 0$$

If A squared, Eigenvectors same, Eigenvalues squared

$$\begin{bmatrix} -3 & 3 \\ 4 & 4 \end{bmatrix} \vec{a} = 0$$

$$a_1 + a_2 = 0$$

$$a_1 = -a_2$$

$$a_2 = -a_1$$

$$\begin{bmatrix} 1 \\ -1 \end{bmatrix} a_1$$

$$\begin{bmatrix} -4 & -3 \\ 4 & 3 \end{bmatrix} \vec{a} = 0$$

$$-4a_1 - 3a_2 = 0$$

$$-4a_1 = 3a_2$$

$$a_1 = -\frac{3}{4}a_2$$

$$\begin{bmatrix} -\frac{3}{4} \\ 1 \end{bmatrix} a_2$$